

APPLICATION NOTE

**MIPS R4000 Slew Rate Control Logic
for Output Buffers**

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One of the largest causes of variation in voltage or noise on the GND and VCC supplies is the switching of the Output Buffers. This variation in supply voltages has two causes:

- the large number of buffers that can switch at the same time
- each buffer driving a large external load

Noise can be calculated using the formula

$$V = L(\Delta i/\Delta t)$$

where V is the noise on the supplies, L is the inductance of the supply pin, bonding wire, etc., and $\Delta i/\Delta t$ is the rate at which the current changes; this term is directly related to how fast the load capacitor is charged or discharged by the output drivers of the microprocessor.

The above variation or noise on the GND and VCC supplies can be reduced by increasing the rise and fall times of the output drive. Although this reduces the $\Delta i/\Delta t$ term, it also increases the propagation delay time of the output signal to its destination and care must be taken to ensure the resulting delay meets system design requirements.

Slew Rate Control for the Output Buffers in R4000

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The R4000 has Slew Rate Control Logic with a negative feedback loop that adjusts the output buffer drive (or slew rate) so that the output buffer is fast enough to meet the system's speed requirement, and no more, in order to minimize the noise for that particular system. As shown in Figure 1, the feedback loop includes path (A) from pins IO_Out to IO_In . This path represents the signal path with the longest delay from the R4000 to any other external register which matches the load and the transmission line effect on the PCB (see Reference 1). The Slew Rate Control Logic drives a signal through the IO_Out output buffer and, using a selectable clock, samples the return signal on the IO_In pin. If the signal arrives earlier than the maximum acceptable delay, which is selected by the designer, the Slew Rate Control Logic slows the speed of the IO_Out output buffer; if the signal arrives later, it speeds up the IO_Out output buffer.

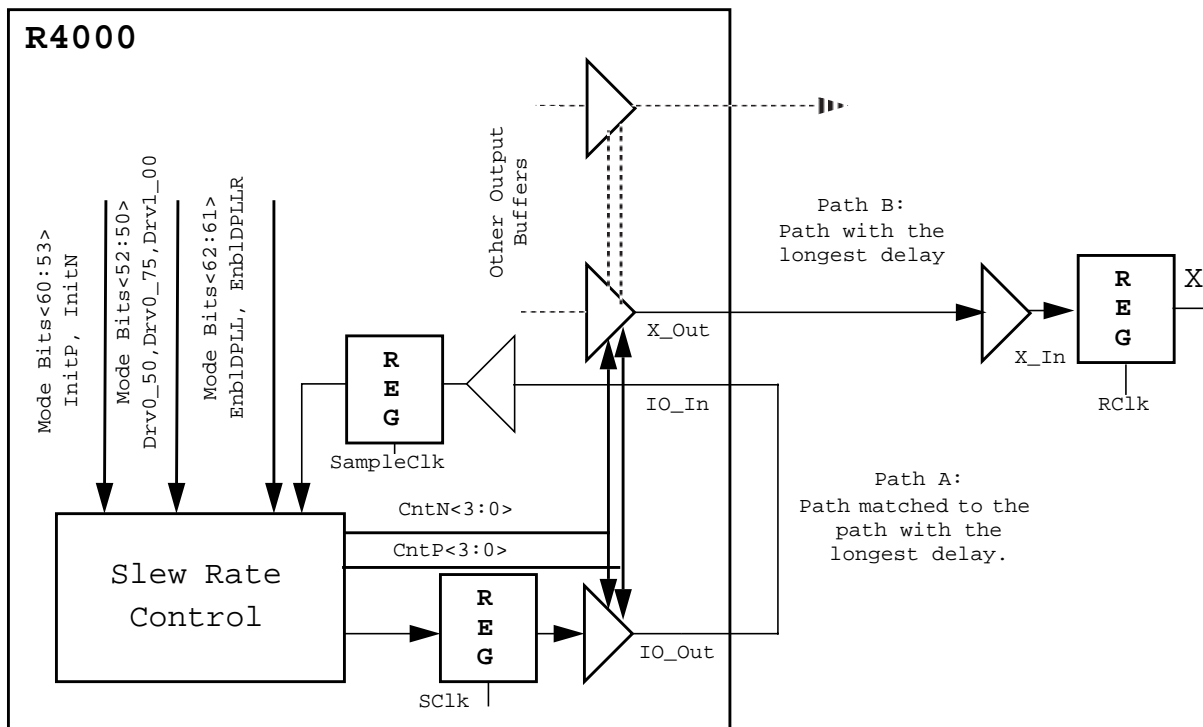


Figure 1: Overview of the Slew Rate Control Mechanism in R4000

The Slew Rate Control Logic uses the signals $CntN<3:0>$ and $CntP<3:0>$ to control the N-channel and the P-channel devices, respectively, of the predrivers to the final stage of the output buffers. This allows $CntN<3:0>$ signals to adjust the rising edge of the outputs and $CntP<3:0>$ to adjust the falling edge, giving 16 independent states of control in each direction. All ones on $CntN<3:0>$ provides the fastest pullup rate and all zeros on $CntP<3:0>$ provides the fastest pulldown rate. The variation in the least significant bit (LSB), $CntN<0>$ or $CntP<0>$, provides the smallest resolution. Note that there is always some jitter at the output due to the fact that the control mechanism is continuously trying to converge to the programmed maximum adjustable delay value by switching the state of the LSBs $CntN<0>$ and $CntP<0>$. The signals, $CntN<3:0>$ and $CntP<3:0>$, that control the output buffer for IO_Out pin also control the output buffers for the rest of the pins of the R4000. Thus, the slew rate of the output buffers will vary in the same manner as the variation of the IO_OUT buffer; however, the amount of variation will depend on the external loading on each pin.

The R4000 has three modes for controlling the slew rates of the output buffers and in turn minimizing the noise caused by $\Delta i/\Delta t$. As shown in Table 1, these three modes are selected by Mode Bits<62:61>, Enb1DPLL and Enb1DPLLr, respectively.

Table 1: Modes of Operation for Controlling the Slew Rate

EnbIDPLL	EnbIDPLLr	Modes of Operation
0	0	Keep the slew rate of the output drivers fixed as set by the Mode Bits<60:57> and <56:53>, programmed during boot time .
0	1	Enable the mechanism to dynamically control the slew rate only during ColdReset and thereafter retain the slew rate of the output drivers.
1	X (don't care)	Enable the mechanism to dynamically control the slew rate during ColdReset and Normal Operation .

The first mode in Table 1, which keeps the slew rate of the output buffer fixed, is useful for the situation in which any jitter associated with the operation of the slew rate control mechanism cannot be tolerated, and the variation in temperature and supply voltage after ColdReset is expected to be small.

To dynamically control the slew rate, either during the ColdReset operation or during both ColdReset and Normal operations, the R4000 does the following.

1. It uses Mode Bits<60:57> and <56:53> (also referred as InitP and InitN, respectively), programmed during the boot time, to set the initial values of CntN<3:0> and CntP<3:0>, which in turn sets the initial slew rate of the output buffers.
2. It uses the “drive-off” delay, programmed at boot time through Mode Bits <52:50> (Drv1_00, Drv0_75 or Drv0_50), as the maximum acceptable delay of the data and, accordingly, selects the SampleClk to sample the IO_In as shown in Figure 1. Table 2 shows the truth table for Mode Bits<52:50>; the remaining combinations are reserved
3. It uses the path between IO_Out pin and IO_In pin, as described earlier.

Table 2: The Truth Table for Mode Bits <52:50>, Setting Drive-off Delays

Mode	Drv1_00	Drv0_75	Drv_5	Drive-off time
Dynamic	0	0	1	$0.50 * T^{(2)}$
Dynamic	0	1	0	$0.75 * T^{(2)}$
Dynamic	1	0	0	$1.00 * T^{(2)}$
Fixed	X	X	X	Note 1

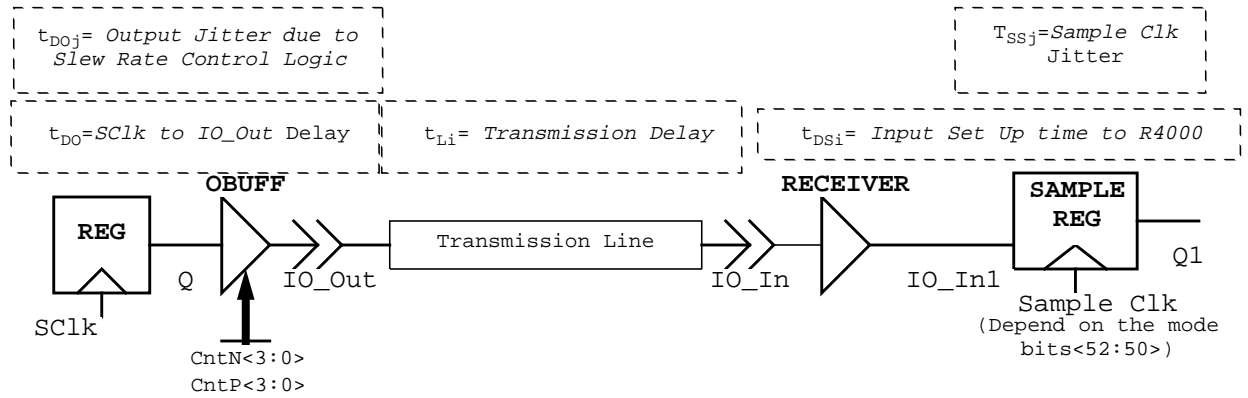
Notes: 1) The Fixed “Drive-off” Time Is Selected By Initn and Initp (Mode Bits<60:53>).
 2) ‘T’ is the period of the Master clock (MasterClk).

To make the programmed delay value independent of system clock frequency (MasterClk), the delay value is specified in terms of a percentage of one cycle time instead of a constant. Also, note that this mechanism is not affected by the interface clocks divisor (Mode Bits<49:47>), it is always a fraction of one (MasterClk) cycle.

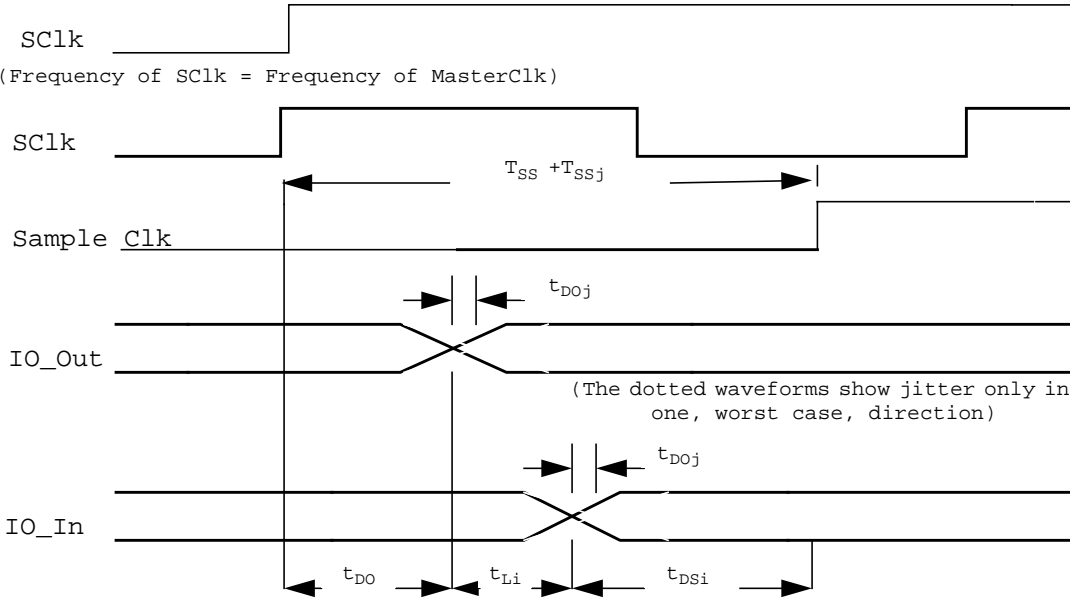
As mentioned in the previous section, for the dynamic control of the output buffer slew rate, R4000 uses the “drive off delay” programmed by Mode Bits<52:50> during boot time. This section explains how to determine the setting of these mode bits. In general, a path is selected (Path B in Figure 1) with the longest delay from the output pin to the output of the receiving data register (node X in Figure 1). Then, the transmission line effect and the capacitive loading of the IO_Out is matched with the transmission line effect and the capacitive loading of Path B (see Reference 1 for detail). Finally, after calculating at what time the SampleClk in the Path A should be, the appropriate Mode Bits<52:50> (Drv1_00, Drv0_75 or Drv0_50) are selected.

Figure 2 shows the Path A again and the corresponding timing diagram with all the delay components labelled. The equation of the delay T_{SS} plus the clock jitter T_{SSj} in terms of the output buffer delay (t_{DO}) is also shown (Equation 1). T_{SS} is the time from the rising edge of the SClk, which triggers the IO_Out, to the SampleClk edge which latches the IO_In at a selected time. T_{SSj} is the jitter of the SampleClk with reference to the SClk. The T_{SSj} could be both positive or negative; however, a positive number is chosen because, in this case its affect is worse than the affect of a negative number. t_{DOj} is the jitter on the output caused by the slew rate control logic which is continuously trying to converge to the programmed “drive off” delay value by switching the state of the LSBs, CntN<0> and CntP<0>.

Figure 3 shows Path B again and corresponding timing diagram with all the delay components labelled. This path is an example of a signal path from the R4000 to a receive-data register. This register could be part of external agent chip or an individual component on the PCB. The rising edge of SClk drives the data, and the rising edge of the RClkD latches the data into the register. T_{SR} is the delay between the RClk and the SClk, T_{SRD} is delay between the RClkD and the SClk and T_{SRj} is the jitter on RClk or RClkD with respect to SClk. The T_{SRDj} could be a positive or a negative number; however, a negative number chosen because, in this case, its affect is worse than the affect of a positive number. The equation of the delay time T_{SRD} is shown in Equation 2 in Figure 3.



(Frequency of SClk < Frequency of MasterClk)



$$Eq\ 1: T_{SS} + T_{SSj} = t_{D0} + t_{Li} + t_{DSi}$$

Figure 2: Path from pins IO_Out to IO_In.

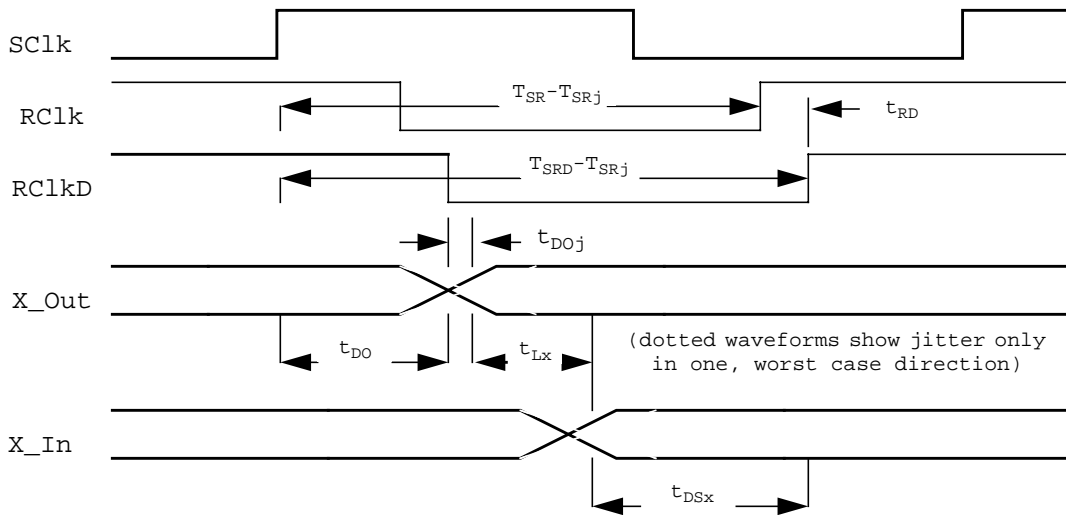
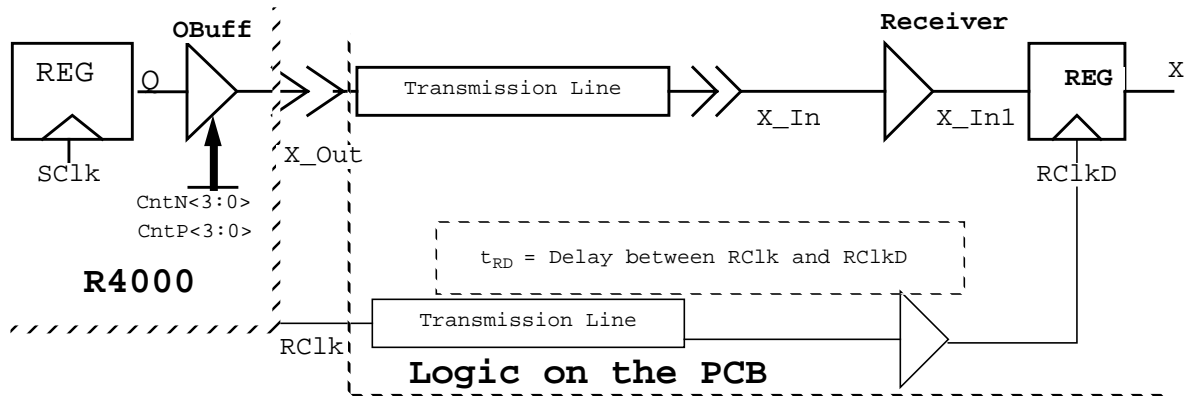
t_{DOj} = Output Jitter due to Edge Control Logic

T_{SRj} = Clock Jitter

t_{DO} = SClk to X_Out Delay

t_{Lx} = Transmission Delay

t_{DSx} = Input Set Up time to ext.logic



Eq 2: $T_{SRD} - T_{SRj} = t_{DO} + t_{DOj} + t_{Lx} + t_{DSx}$

Figure 3: Representative Worst Case Path from R4000 to External Component

If the capacitive load and the transmission line between the pins IO_Out and IO_In are matched with the loading and the transmission line between X_Out and X_In, then the output buffer delay, t_{DO} , is equal in both the paths, and $t_{Li} = t_{Lx}$. Thus, deriving t_{DO} from the Equation 2 and substituting into the equation Equation 1, T_{SS} can be derived to decide which of the Mode Bits Drv1_00, Drv0_75, or Drv0_50 needs to be selected.

$$Eq\ 3: DrvX_{XX} < T_{SS} = T_{SRD} - T_{SRj} - t_{SSj} + t_{DSi} - t_{DSx} - t_{DOj}$$

The following examples show how to set Mode Bits <52:50>.

Example 1

In a system with SysCkRatio (Mode bits<17:15>) set to 0 (Ref: 1) , the frequency of RClk is equal to the frequency of MasterClk . If the delay, equal to the delay between RClk and RClkD , is added between SyncIn pin and SyncOut pin, the RClkD is locked to the MasterClk . But since RClkD leads SCLk and MasterClk by $0.25*T$; this means, T_{SRD} is $(0.75*T)$.

The R4000 AC specification (Ref.1) shows clock jitter = 0.5 ns, data set up time = 5.0 ns, and $t_{DOj} = 1$ ns. Assuming $t_{DSX} = 7$ ns, Equation 3 can be rewritten as:

$$Eq\ 4: T_{SS} = (0.75*T) - 0.5ns - 0.5ns + 5ns - 7ns - 1ns = 0.75*T - 4ns$$

Thus, for a 50 MHz system, T_{SS} would be 11 ns; so Drv0_50 would be set to 1.

Example 2

In a system with SysCkRatio (Mode bits<17:15>) set to 0, the frequency of the RClk is equal to the frequency of MasterClk . If the Sync_In pin is shorted to the SyncOut pin, the RClk is locked to the MasterClk . Since RClk leads SCLk and MasterClk by $0.25*T$; this means, T_{SRD} is equal to $(0.75*T + t_{RD})$, where t_{RD} is the delay between RClk and RClkD . Thus, Equation 4, from Example 1, can be rewritten as,

$$T_{SS} = (0.75*T + t_{RD}) - 4ns$$

If $t_{RD} > 4ns$ then T_{SRD} could be $> (0.75*T+4ns)$, making $T_{SS} > 0.75*T$; in which case Drv0_75 is set to 1.

Example 3

In a system with `SysCkRatio` set to 1, the frequency of `RClk` is two-thirds the frequency of `MasterClk`, so T_{SRD} is equal to $(1.0 \cdot T)$ given in Example 1, and T_{SRD} is greater than $(1.0 \cdot T)$ given in Example 2.

Thus, in the first example `Drv0_75` is set to 1, and in the second example `Drv1_00` is set to 1.

Example 4

Similarly, in the system with the `SysCkRatio` set to 2, the frequency of `RClk` is half the frequency of `MasterClk`; so `Drv1_00` is set to 1 in Examples 1 and 2.

The R4000 has Slew Rate Control Logic with negative feedback loop that adjusts the output buffer drive so that the output buffer is only fast enough to meet the system's speed requirement, and no more, in order to minimize the noise for that particular system. The maximum acceptable delay is programmed at the boot time using the "drive-off" Mode Bits<52:50> (Drv1_00 , Drv0_75 or Drv0_50). It also provides the flexibility to select whether to enable the control only during the ColdReset time, or during the ColdReset time and the Normal operation, or to select a fixed slew rate as programmed by the Mode Bits<60:53> (InitP and InitN). Some examples were shown in this document on how to select the "drive-off" delay mode bits.

References

Reference 1: MIPS R4000 Processor Interface Specification

Reference 2: MIPS R4000 Microprocessor User's Manual

Reader's Comments

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